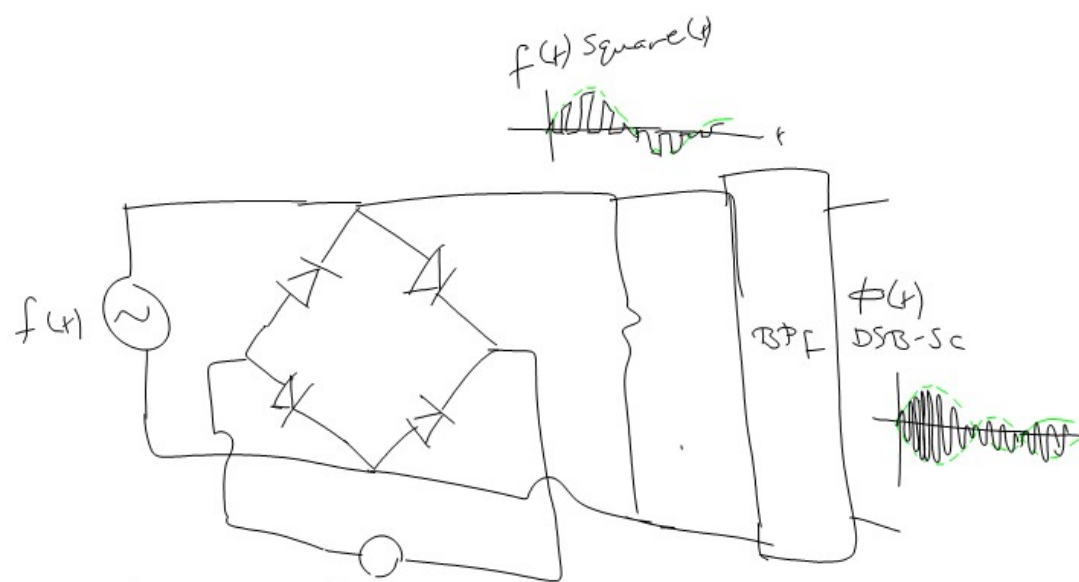




- * Note that $\cos \omega_c t$
the amplitude of the carrier must be greater than the message signal

Use of non linear device

- * We can perform the mixing process by using some non linear devices such as diodes
- * The nonlinearity of the diode can be understood from the IV characteristics
- $$I(V) = I_s \left(1 - e^{-\frac{V}{kTq}} \right)$$



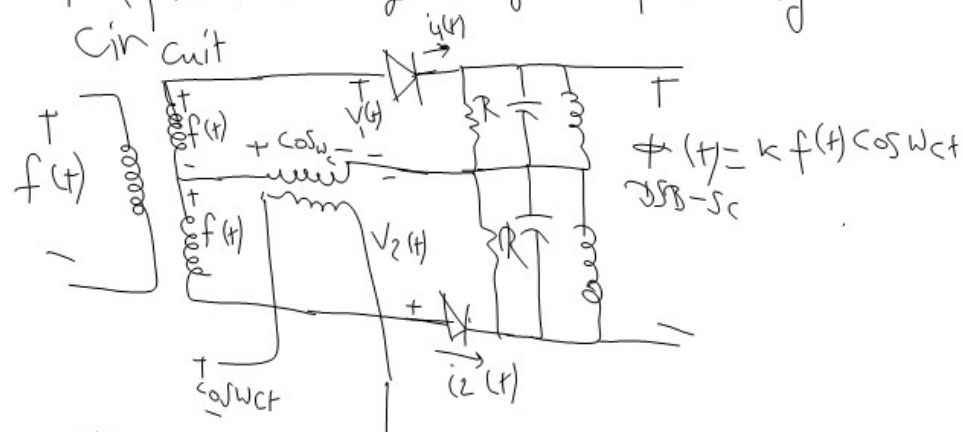
- * The nonlinearity can be approximated by using the power series expansion for $I(V)$

$$I(V) = a_1 V + a_2 V^2 + a_3 V^3 + a_4 V^4 + \dots$$

Note that for small voltages the higher order terms can be neglected

- * Higher order terms are $a_3 V^3 + a_4 V^4 + \dots$

- * The nonlinear mixing action can be implemented by using the following circuit



$$V_1(t) = \cos \omega_c t + f(t)$$

$$V_2(t) = \cos \omega_c t - f(t)$$

$$i_1(t) = a_1 V_1(t) + a_2 V_1^2(t)$$

$$i_2(t) = a_1 V_2(t) + a_2 V_2^2(t)$$

The voltage at the resistors is given by

$$\begin{aligned} V_0(t) &= R i_1(t) - R i_2(t) \\ &= R [a_1 V_1(t) + a_2 V_1^2(t) - a_1 V_2(t) - a_2 V_2^2(t)] \\ &= R [a_1 (V_1(t) - V_2(t)) + a_2 (V_1^2(t) - V_2^2(t))] \\ &= R [a_1 2f(t) + a_2 4f(t) \cos \omega_c t] \end{aligned}$$

- * The signal measure at the BPF output is given by
- $$\phi(t) = 4a_2 R f(t) \cos \omega_c t$$

5.1.2 Demodulation of DSB-SC

* In order to recover $f(t)$ from $\phi(t)$ we just multiply $\phi(t)$ by the carrier. However the Locally generated carrier in the receiver must be synchronized with carrier in the transmitter

Synchronization can be achieved sending a sine wave whose frequency and phase are related to the carrier in the receiver

* This sine wave is called a pilot signal

